

UNIT 3: Advanced Control System

Chp 15: feed forward + Ratio Control

Feed back

(+) Corrective action occurs for any deviation input (regardless of source)

Requires minimal knowledge about sys

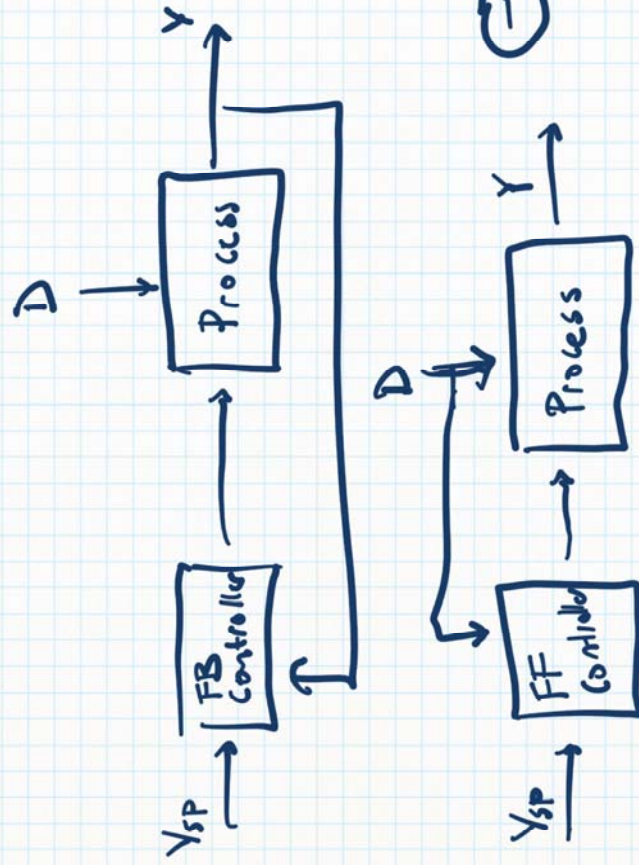
PID controllers are robust

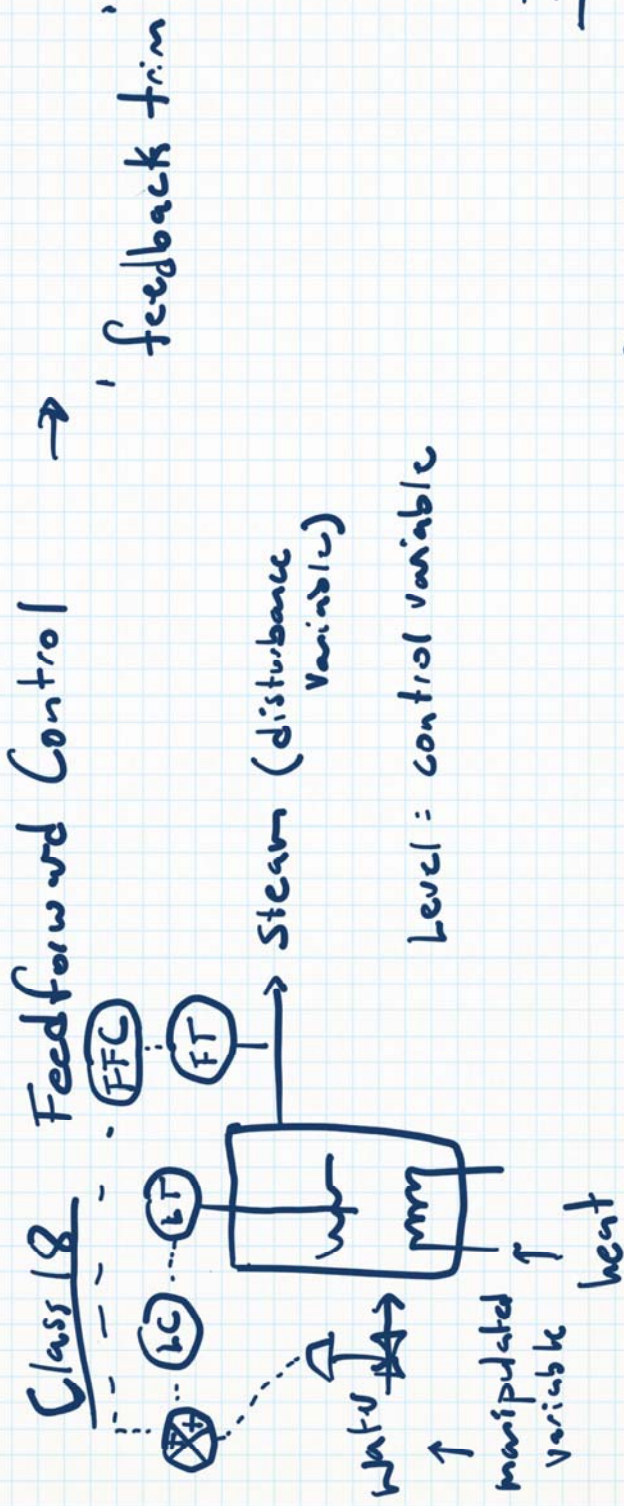
(-) A perfect control not possible

Cannot compensate for known disturbances before they happen

Not good for sys w/ large time constants

CV is hard/impossible to measure online





Example Systems

□ Blending operation

□ flow rates □ Stoichiometry in Reactors

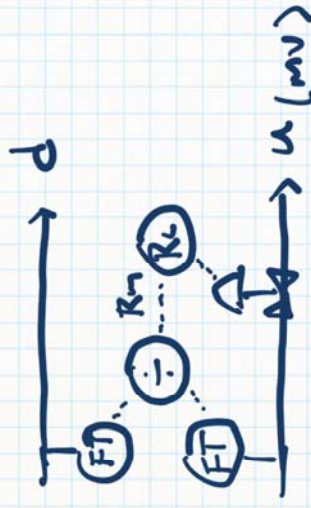
□ Reflux in Distillations

□ furnace = fuel/air ratio

"Ratio Control" ← subset of feed forward

$$R = \frac{u}{d}$$

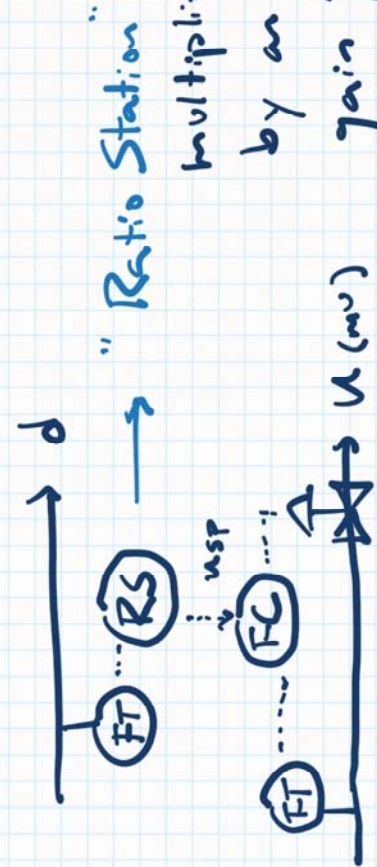
← manipulated variable ← disturbance variable



METHOD 1

process gain will change

$$K_p = \left(\frac{R_l}{R_m} \right) \Big|_d = \frac{1}{d}$$



"Ratio Station"

Multiplies signal by an adjustable gain K_R

- output is the setpoint for the controller

METHOD 2

Process Gain is constant

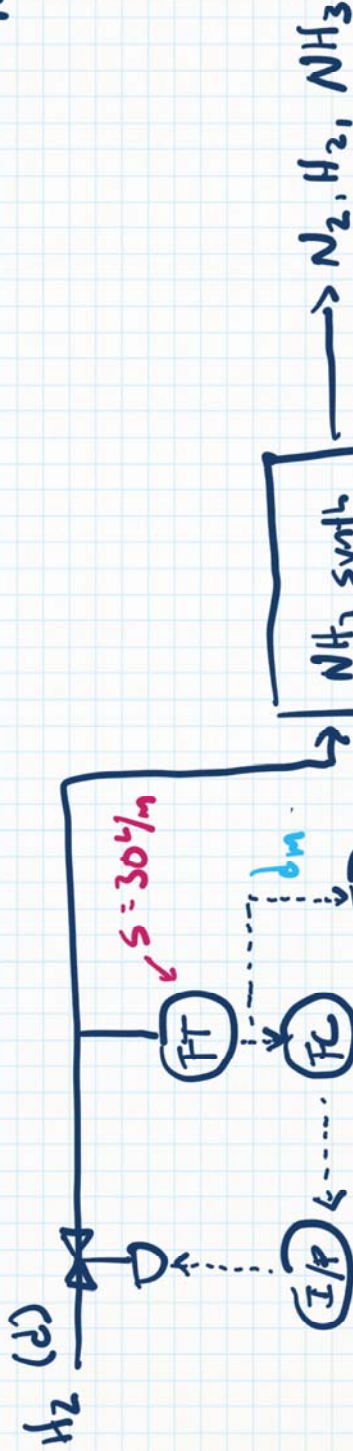
$$K_R = R_d \frac{S_d}{S_u} \left[\text{Spans of Flow transmitters for disturbance variable and measure} \right]$$

Ex] Stoichiometric Control for Ammonia Synthesis



$$\frac{H_2}{N_2} \text{ ratio } 3:1$$

$$R_d = \frac{u}{d} = \frac{1}{3}$$



$$K_R = R_d \frac{S_d}{S_m}$$

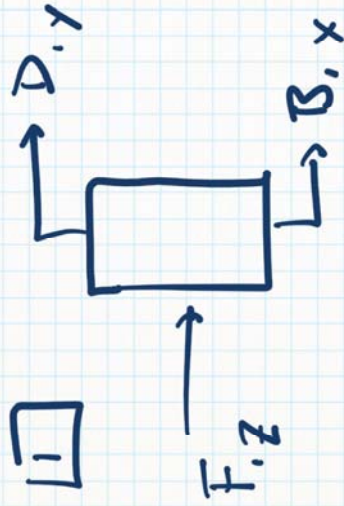
device w/
an adj-stable gain (K_R)

$$K_R = \frac{1}{3} \left(\frac{30\%/min}{15\%/min} \right)$$

$$U_{sp} = K_R d_m \uparrow$$

$$= 2/3$$

Designing FF systems: 1 Steady State 2 Dynamic



$$\bar{F} = \bar{D} + \bar{B}$$

$$\bar{F}\bar{z} = \bar{D}\bar{y} + \bar{B}\bar{x}$$

← solve for \bar{B} and then

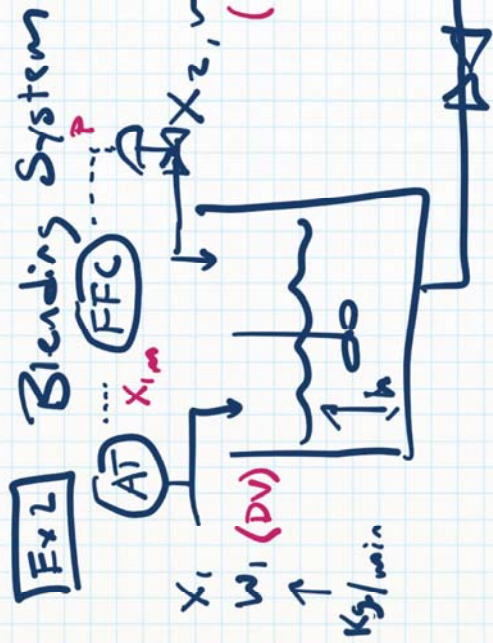
$$\bar{D} = \frac{\bar{F}(\bar{z} - \bar{x})}{\bar{y} - \bar{x}}$$

↑
Hold these at sp

"FFC is continual attempt to balance the material or energy that must be delivered to process against a disturbance"

FF. control law for distillation column

$$D(t) = \frac{F(t) [Z(t) - x_{sp}]}{y_{sp} - x_{sp}}$$



M.B. $\bar{w} = \bar{w}_1 + \bar{w}_2$

C.M.B. $\bar{xw} = \bar{w}_1 \bar{x}_1 + \bar{w}_2 \bar{x}_2$

combine + solve \bar{w}_2 (mV)

$\bar{w}_2 = \frac{\bar{w}_1 (\bar{x} - \bar{x}_1)}{\bar{x}_2 - \bar{x}}$

measure this convert to $-x_{1m}$ (mA)

A Composition Measure

$K_t = \frac{\text{measured signal}}{\text{measured flow rate}}$

$K_t = \frac{20 - 4 \text{ mA}}{S_t}$

S_t span of flow rates

$w = C_V \sqrt{h}$ (v)

B Change this to controller output signal

$= \frac{x_{1m} - 4 \text{ mA}}{x_1(t) - (x_1)_0}$

$x_{1m}(t) = K_t [x_1(t) - x_{10}] + 4$

[B] Pressure Transducer and Control Valve



$$K_{IP} K_V = \frac{\text{flow rate}}{\text{controller output}} = \frac{W_2(t) - W_{2(0)}}{P(t) - 4 \text{ mA}}$$

$$W_2(t) = K_V K_{IP} [P(t) - 4] + (W_2)_0$$

↑
mass balance

Combine mass balance Eq. [MB] w/ [A] and [B] *mA coming off sensor*

$$P(t) = C_1 + C_2 \left[\frac{K_t X_{sp} - X_{1m}(t) + C_3}{\bar{X}_2 - X_{sp}} \right]$$

↑
mA out of controller

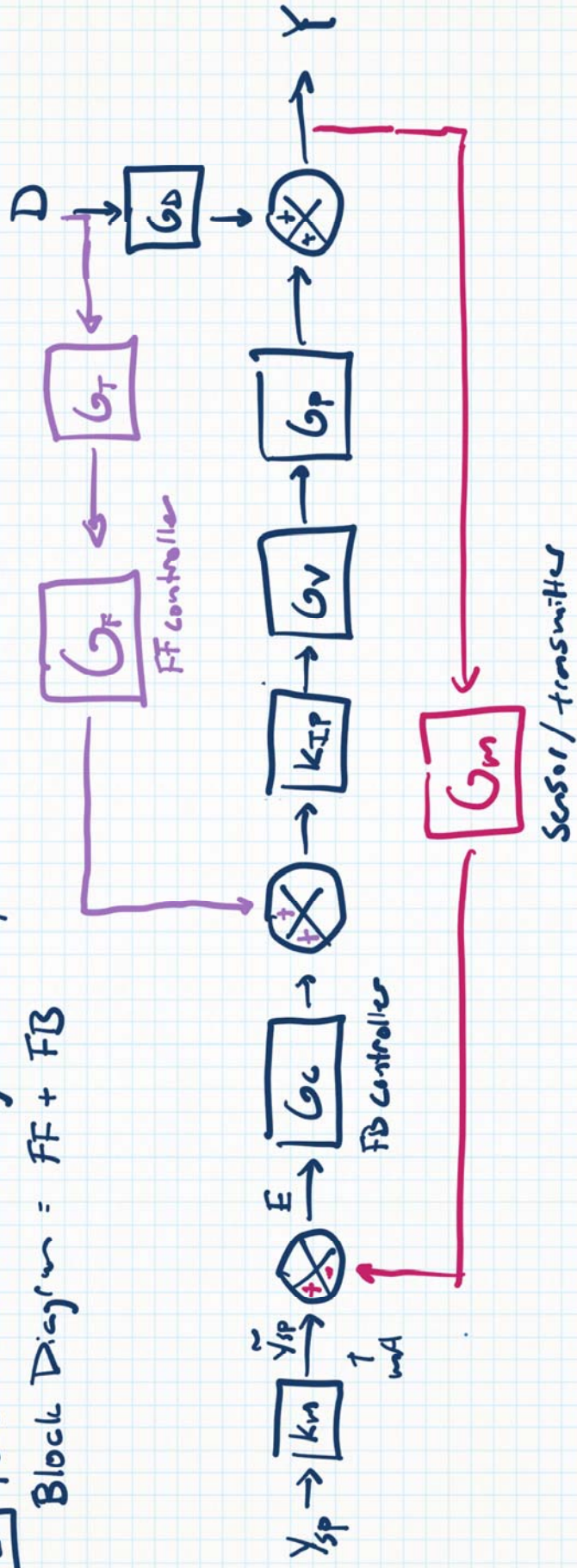
$$C_1 = \frac{4 - W_{20}}{K_V K_{IP}}$$

$$C_2 = \frac{W}{K_V K_{IP} K_t}$$

$$C_3 = 4 + K_t (X_1)_0$$

2 Feed Forward Design = Dynamic Model

Block Diagram = FF + FB



$$\frac{Y(s)}{D(s)} = \frac{G_d + G_r G_f K_{IP} G_v G_p}{1 + G_c K_{IP} G_v G_p G_m}$$



CLASS 20: Feed Forward Control (w/ Dynamic Systems)

$$\frac{Y(s)}{D(s)} = \frac{G_d + G_t G_f G_v G_p}{1 + G_c G_v G_p G_m}$$

Ex 1

$$G_t = K_t$$
$$G_v = K_v$$
$$G_d = \frac{K_d}{T_P s + 1}$$

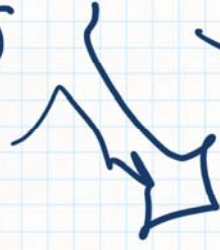
Perfect control $Y(s) = 0$

$$G_d + G_t G_f G_v G_p = 0$$

$$G_f = - \frac{G_d}{G_t G_v G_p}$$

Feed forward Controller TF.

$$G_p = \frac{K_p}{T_P s + 1}$$



$$G_f = - \frac{K_d}{(K_t K_v K_p)} \left(\frac{T_P s + 1}{T_D s + 1} \right)$$

* Lead-Lag TF. * Lead * Lag

Ex 2

$$G_d = \frac{K_d}{T_d \cdot s + 1}$$

$$G_p = \frac{K_p C^{-\theta s}}{T_p \cdot s + 1}$$

Not physically realizable

$$G_f = - \left(\frac{K_d}{K_t K_v K_p} \right) \left(\frac{T_p \cdot s + 1}{T_d \cdot s + 1} \right) e^{+\theta s}$$

handle this

negative time delay
by lumping θ w/ T_p

$$T_p + \theta = \text{New lead time constant}$$

Ex 3

$$G_d = \frac{K_d}{T_d \cdot s + 1}$$

$$G_p = \frac{K_p}{(T_{p1} \cdot s + 1)(T_{p2} \cdot s + 1)}$$

$$G_f = - \left(\frac{K_d}{K_t K_v K_p} \right) \left(\frac{(T_{p1} \cdot s + 1)(T_{p2} \cdot s + 1)}{(T_{p2} \cdot s + 1)} \right)$$

Also Not

physically realizable

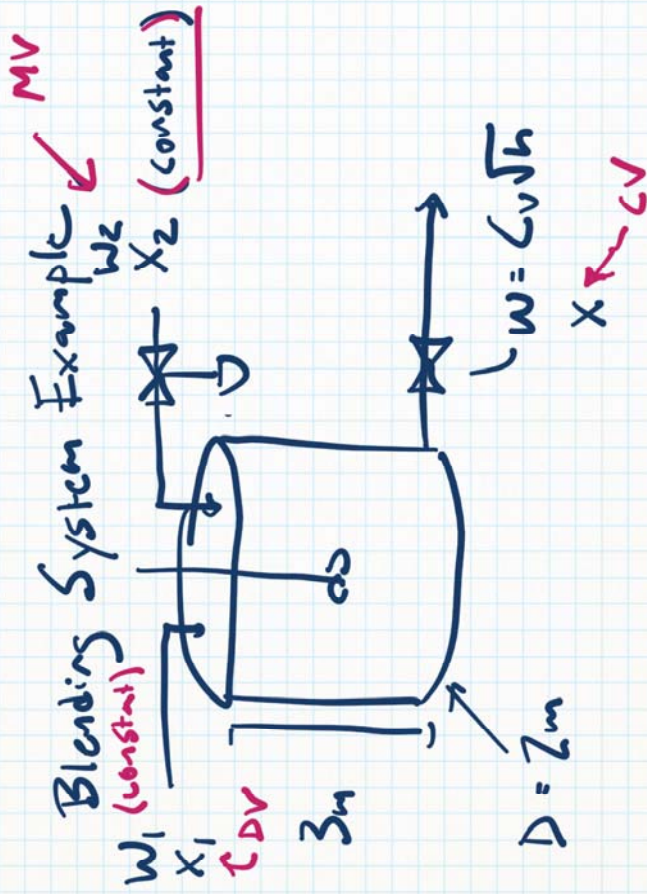
$$T_{p1} + T_{p2} = \text{New lead time constant}$$

Stability = G_f does NOT appear in the characteristic eqn. for the way it is typically connected (where both controller (FB+FF) outputs are summed)

↳ Thus, it does NOT affect stability ...

Typical Form of FFC

$$G_f(s) = \frac{U(s)}{D(s)} = \frac{K_f (\tau_1 s + 1) e^{-\theta_s}}{(\tau_2 s + 1)}$$



Nominal Steady State

$$\bar{W}_1 = 650 \frac{\text{kg}}{\text{min}} \quad \bar{X}_1 = 0.2$$

$$\bar{h} = 1.5m$$

$$\bar{W}_2 = 350 \frac{\text{kg}}{\text{min}} \quad \bar{X}_2 = 0.6$$

$$\rho = 13/\text{cm}^3$$

$$\bar{X} = 0.34$$

Using derivation from class 19 we get the following

$$P(1) = 4 + 1.083 \left[\frac{3Z X_{sp} - X_{1,m}(+) + 4}{0.6 - X_{sp}} \right]$$

With Dynamics:

$$G_f = \frac{G_d}{K_{IP} G_e G_v G_T}$$

$$G_P(s) = \frac{K_P}{T_s + 1}$$

$$T = \frac{V_P}{\bar{W}} \leftarrow \pi R^2 h$$

Chp. 4 (derived Tank problem)

where we have $(P \neq K_f)$

$$K_P = \frac{X_2 - X}{\bar{W}}$$

$$K_f = \frac{\bar{W}}{W}$$



$$G_f = \frac{-6d}{K_{IP} G_t G_v G_P}$$

$$= -\frac{0.65}{4.71s+1}$$

Assume PI feedback controller

$$K_c = 21.9$$

$$T_I = 3.33 \text{ min}$$

$$G_c = K_c \left(1 + \frac{1}{T_I s} \right)$$

$$(0.75)(32e^{-s}) \left(\frac{25}{0.0833 \cdot s + 1} \right) \left(\frac{2.6 \times 10^{-4}}{(4.71s+1)} \right)$$

$$= -4.17 (0.0833 \cdot s + 1) e^{-s}$$

G_f

$$= \frac{-4.17 (1.0833 \cdot s + 1)}{\alpha (1.0833) \cdot s + 1}$$

$$\text{where } \alpha = 1$$

1] disregard our negative time delay

2] Add 0 to lead time constant

3] Make ~ lag time constant

that is 10% of our new lead constant

Use 'impz' function in Matlab to model response

FF by itself is not good. Include 'Feedback Trim'

- modeling errors
- unmeasured disturbances

Tuning (process after you have best guess values for G_f)

$$G_f = \frac{K_f (\tau_1 s + 1) e^{-\theta_s}}{(\tau_2 s + 1)}$$

Step 1 set τ_1 and τ_2 to 200 to check gain

→ increase disturbance variable 3-5%

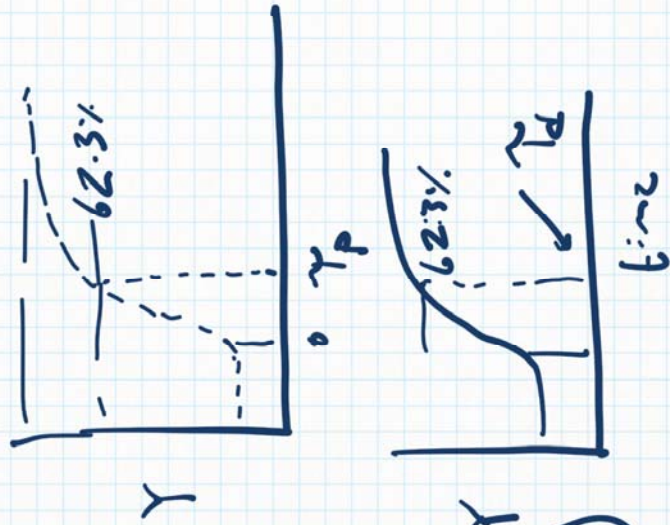
→ Goal: minimize offset

Step 2 τ_1 and τ_2 use process data for first guess

(fit of exp. data)

$$\tau_1: \tau_p \text{ (less)}$$

$$\tau_2: \tau_D \text{ (larger)}$$



Servo Step

disturbance Step

(Set point const. disturbance) Step

See this before (open loop step test)

$$K_p = \frac{Y_{\infty} - Y_0}{\Delta u} \leftarrow \text{magnitude of step}$$

set point

$$K_d = \frac{Y_{\infty} - Y_0}{\Delta d} \leftarrow \text{magnitude of disturbance}$$

step

No Data? SAD ...

$$\frac{\tau_1}{\tau_2} = 2$$

$$\frac{\tau_1}{\tau_2} = 0.5$$

Step 3

Fine tuning

$$\tau_1 \rightarrow \tau_2$$

Class 22 (The FINAL Lecture) Cascade Control, (and some other Advanced Techniques)

Enhanced Disturbance Rejection

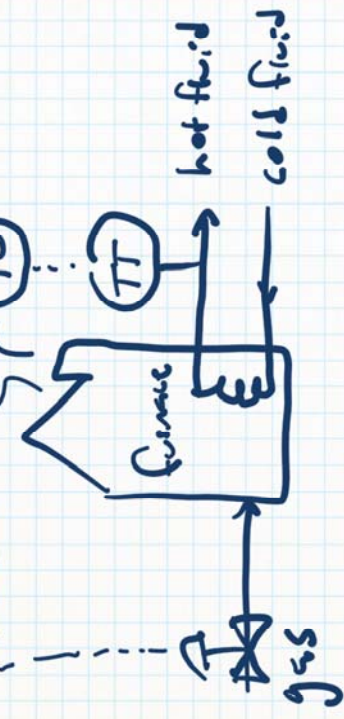
- ① Feed Forward Control w/ FB Trim
 - (+) directly measures disturbance
 - (-) Accurate model is needed, additional sensor
 - (-) controls for ONE SPECIFIC disturbance

② Cascade Control

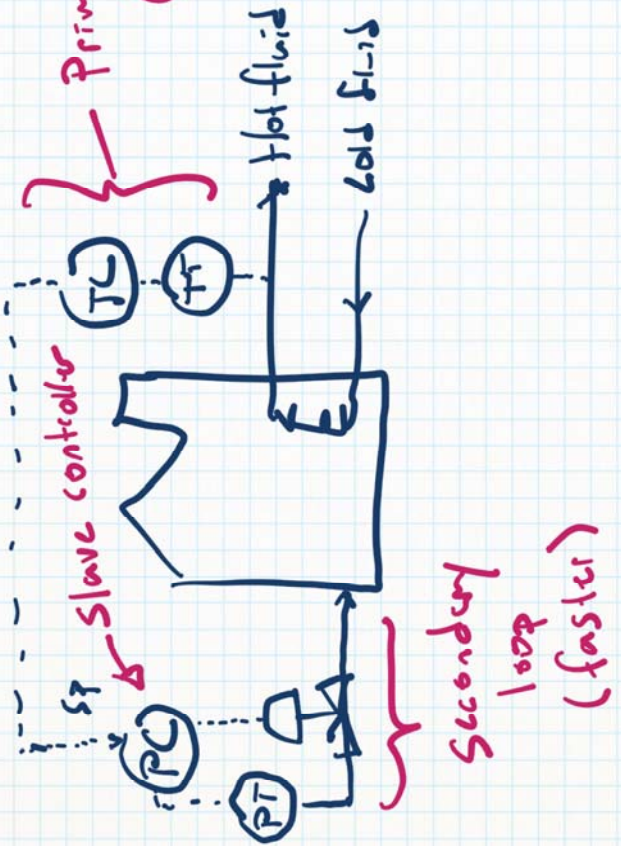
- 2 sensors, 2 controllers, 1 final control element
- 2 nested feedback loops
- identify secondary disturbance variable that can be controlled quickly
- Try this first, if it fails go to FFC w/ FB

Objective: ! Disturbance Rejection

Works well if disturbance is in the hot fluid stream



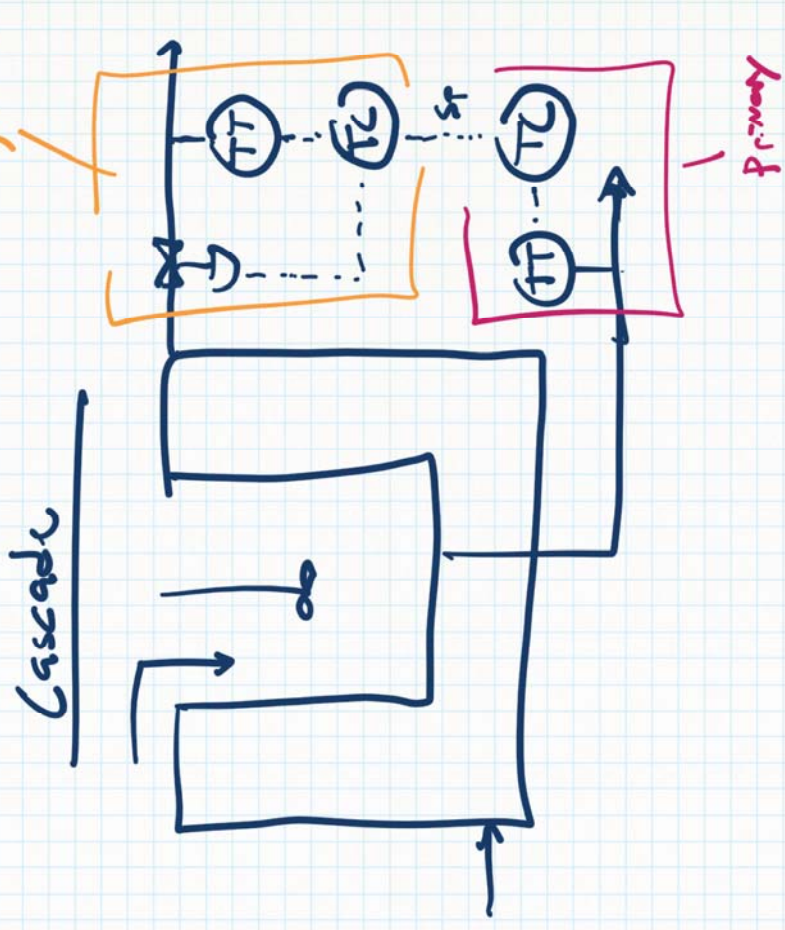
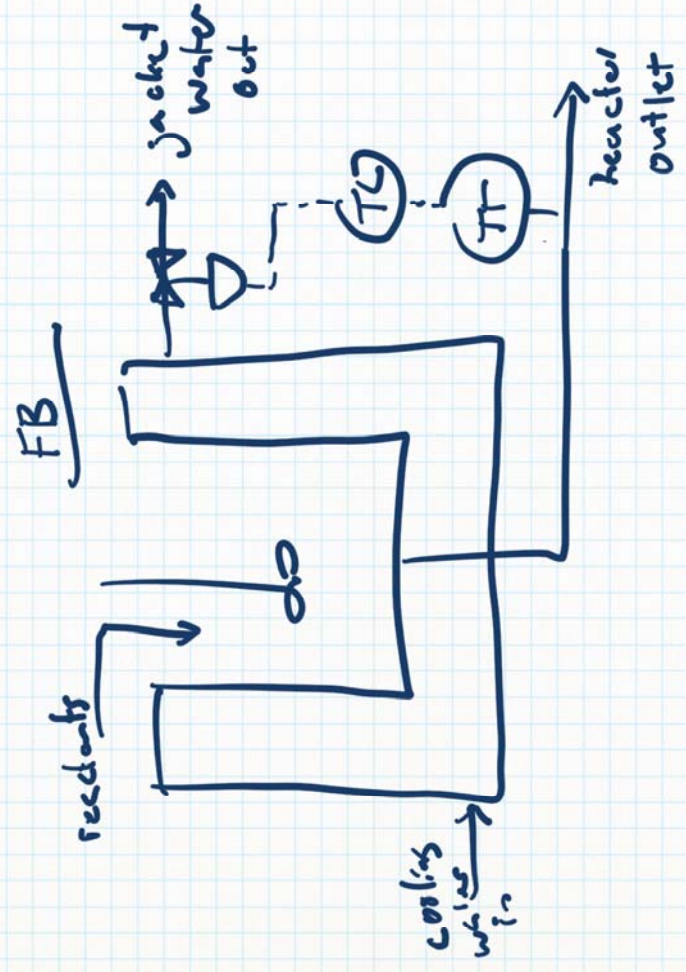
What if we have disturbance here?



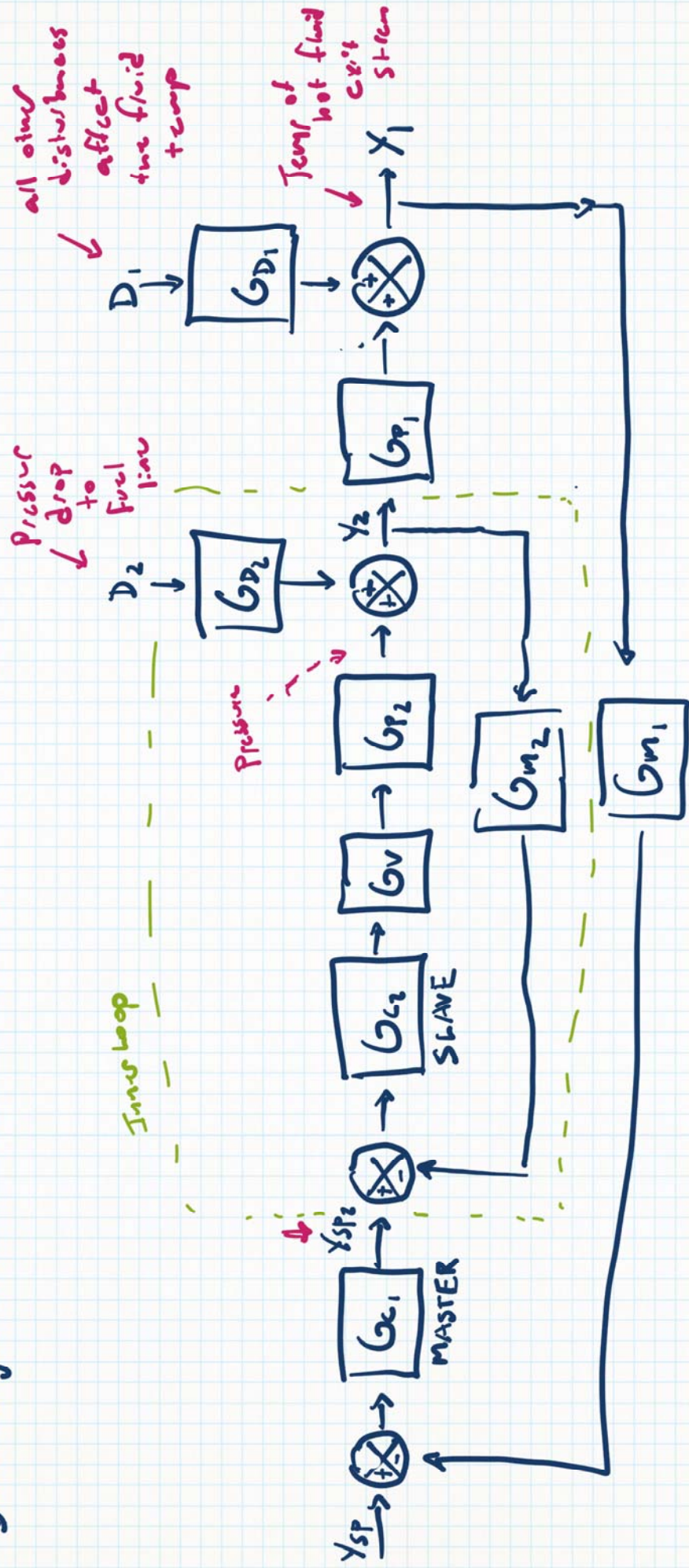
Findings

- ① 2nd variable must be measurable
- ② final control element must have an effect on 1st + 2nd variable
- ③ 2nd variable must be 'inside' the first variable
- ④ Settling time for the inner loop (2nd) must be significantly faster than outer
- ⑤ Output from Master controller is S.P. for the slave...

Ex 2. Cascade control
 Reactor w/ heated jacket



Block Diagram Cascade Control (units from Furnace Ex.)



Transfer functions for Cascade Block diagram

Inner Loop:

$$\frac{Y_1}{D_2} = \frac{G_{P_1} G_{D_2}}{1 + G_{P_2} G_V G_{C_2} G_{M_2} + G_{P_1} G_{P_2} G_V G_{C_2} G_{C_1} G_{M_1}}$$

Set point

$$\frac{Y_2}{Y_{SP_2}} = \frac{G_{C_2} G_V G_{P_2}}{1 + G_{C_2} G_V G_{P_2} G_{M_2}}$$

Outer Loop
whole system

disturbance

$$\frac{Y_1}{D_1} = \frac{G_{D_1} (1 + G_{C_1} G_V G_{P_2} G_{M_2})}{1 + G_{C_2} G_V G_{P_2} G_{M_2} + G_{C_1} G_{C_2} G_V G_{P_2} G_{P_1} G_{M_1}}$$

Set point

$$\frac{Y_1}{Y_{SP_1}} = \frac{G_{C_1} G_{C_2} G_V G_{P_1} G_{P_2}}{1 + G_{C_2} G_V G_{P_2} G_{M_2} + G_{C_1} G_{C_2} G_V G_{P_2} G_{P_1} G_{M_1}}$$

Example Problem

$$G_v = \frac{5}{s+1} \quad G_{p1} = \frac{4}{(4s+1)(2s+1)} \quad G_{p2} = 1$$

$$K_{c2} = 4 \quad G_{d2} = 1 \quad G_{m1} = 0.05 \quad G_{m2} = 0.2$$

$$G_{d1} = \frac{1}{3s+1}$$

Setting time
of inner loop
(Time constant)

$$\frac{Y_2}{Y_{sp2}} = \frac{G_{c2} G_v G_{p2}}{1 + G_{c2} G_v G_{p2} G_{m2}} = \frac{(4)(5/s+1)}{(1 + 4(5/s+1))(0.2)} \quad \begin{matrix} \uparrow \\ \uparrow \end{matrix}$$

→ with cascade

compare to F.B. $G_{c2} = 1$, $G_{m2} = 0$

$$\frac{Y_2}{Y_{sp2}} = \frac{1 \cdot G_v G_{p2}}{1 + 0} = \frac{5}{s+1} \quad \uparrow \tau = 1$$

5x
faster
with cascade

K_{c1} ← P only controller, How much higher (aggressive) can we make this gain w/ cascade vs. FB)

$$1 + G_{c2} G_v G_{P2} G_{m2} + G_{c1} G_{c2} G_v G_{P2} G_{P1} G_{m1} = 0$$

$$1 + 4 \left(\frac{s}{s+1} \right) (0.2) + K_{c1} (4) \left(\frac{s}{s+1} \right) \left(\frac{4}{(4s+1)(2s+1)} \right) (0.05) = 0$$

$$1386 - 32 K_{c1} > 0$$

$$1 + \frac{4}{(s+1)} + \frac{4 K_{c1}}{(s+1)(4s+1)(2s+1)} = 0$$

$$K_{c1} < 43.3$$

$$(s+1)(4s+1)(2s+1) + 4(4s+1)(2s+1) + 4K_{c1} = 0$$

Routh

$$8s^3 + 14s^2 + 7s + 1 + 32s^2 + 24s + 4 + 4K_{c1} = 0$$

$$8s^3 + 46s^2 + 31s + (5 + 4K_{c1}) = 0$$

$$5 + 4K_{c1} \geq 0$$

$$K_{c1} \geq -5/4$$

odd	8	31
even	46	$5 + 4K_{c1}$
→	$1386 - 32K_{c1}$	↖ (+)
		$\frac{46}{46}$

Same problem FB only

$$1 + G_{C1} G_V G_{P1} G_{P2} G_{M1} = 0$$

$$1 + K_{C1} \left(\frac{s}{s+1} \right) \left(\frac{4}{(4s+1)(2s+1)} \right) (0.05) = 0$$

$$(s+1)(4s+1)(2s+1) + K_{C1} = 0$$

$$8s^3 + 14s^2 + 7s + (K_{C1} + 1) = 0$$

$$\rightarrow K_{C1} > -1$$

Tuning:

See below

Routh Array

8	7	
14	$K_{C1} + 1$	
$90 - 8K_{C1}$		14

positive

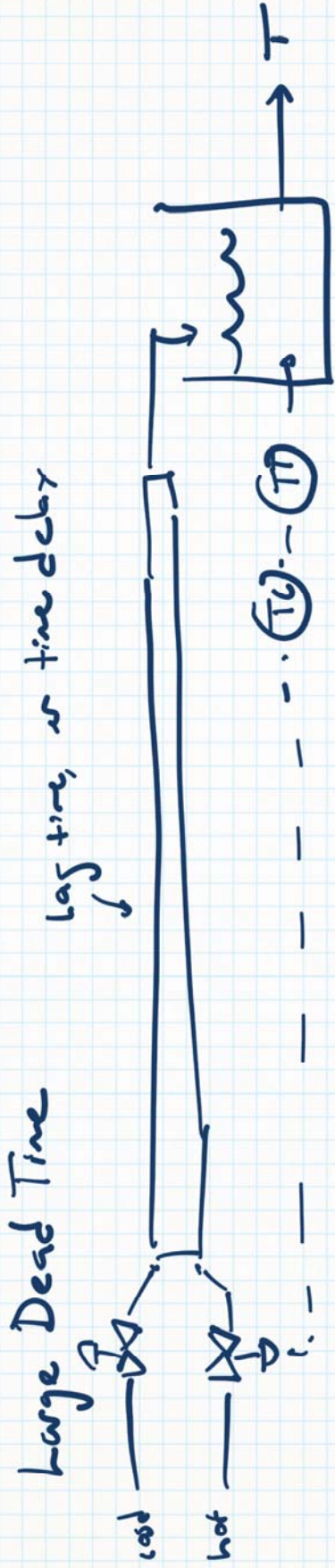
$$90 - 8K_{C1} > 0$$

$$K_{C1} < 11.25$$

↖ w/ Cascade

we can be 4x

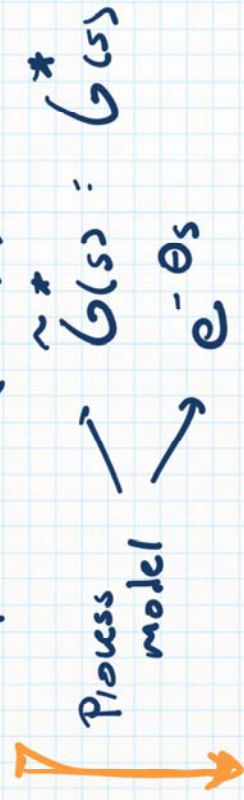
as aggressive



if $\theta \gg \tau$, sensor does not 'see' controller action

if $T < T_{sp}$, valve open, but T doesn't respond for a long time

Smith Predictor (1957)



Model Predictive Control (MPC)

Chp. 20

⚠ - only works well if

you have a ROBUST model

$$\frac{Y}{Y_{sp}} = \frac{G_c G^* \exp(-\theta s)}{1 + G_c G^*}$$

↑ eliminated delay in G.E.

Few other "Advanced" control methods in Chp. 16

▷ Inferential Control = cannot measure your desired CV.

Model \Rightarrow "soft sensor"

take many measurements, use model to predict value of interest

▷ Selective control / override systems ▷ Non-linear

high selector

(Low)

Median selector

split range

override

do not have a constant

controller gain

(K_c will change)

\rightarrow Gain Scheduling \uparrow - controller gains for different error ranges

\rightarrow Fuzzy Logic \rightarrow "Old times controller"

□ Adaptive control → all control parameters (K_p, T_i, T_D)
are adjusted automatically
to compensate for changing process or
environmental conditions

→ Ambient environment (Seasonal, daily)

example: Creepy NEST thermostat

↳ Advanced Topics

Digital control - chp 17

Multivariable control → chp 18 ✓
→ Model predictive control → chp 19